

## EE 209 Homework 6 CMOS Gates

Name: Solutions

Due: See Website

Score: \_\_\_\_\_

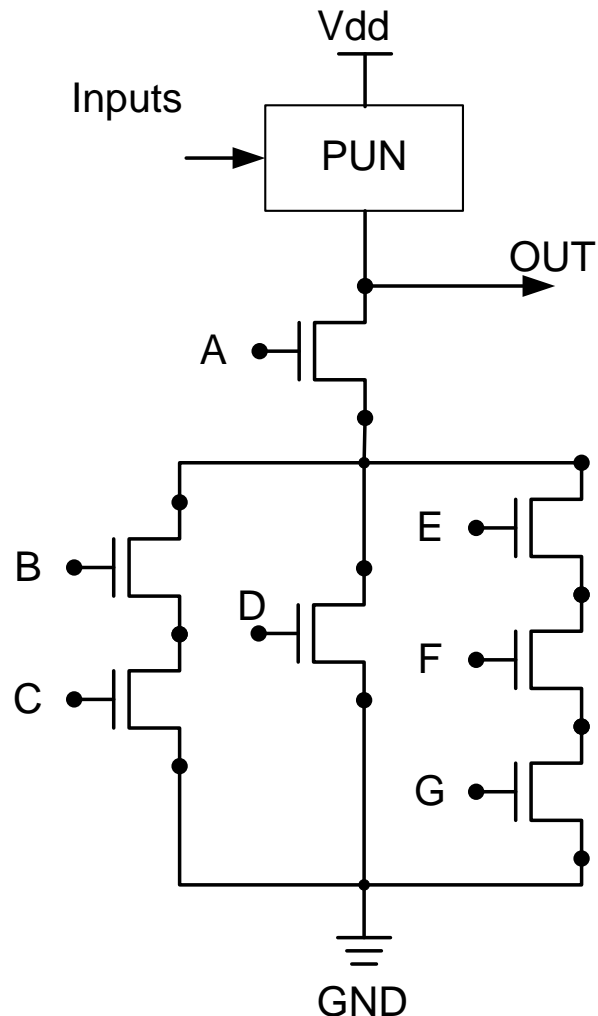
### Paper Submission

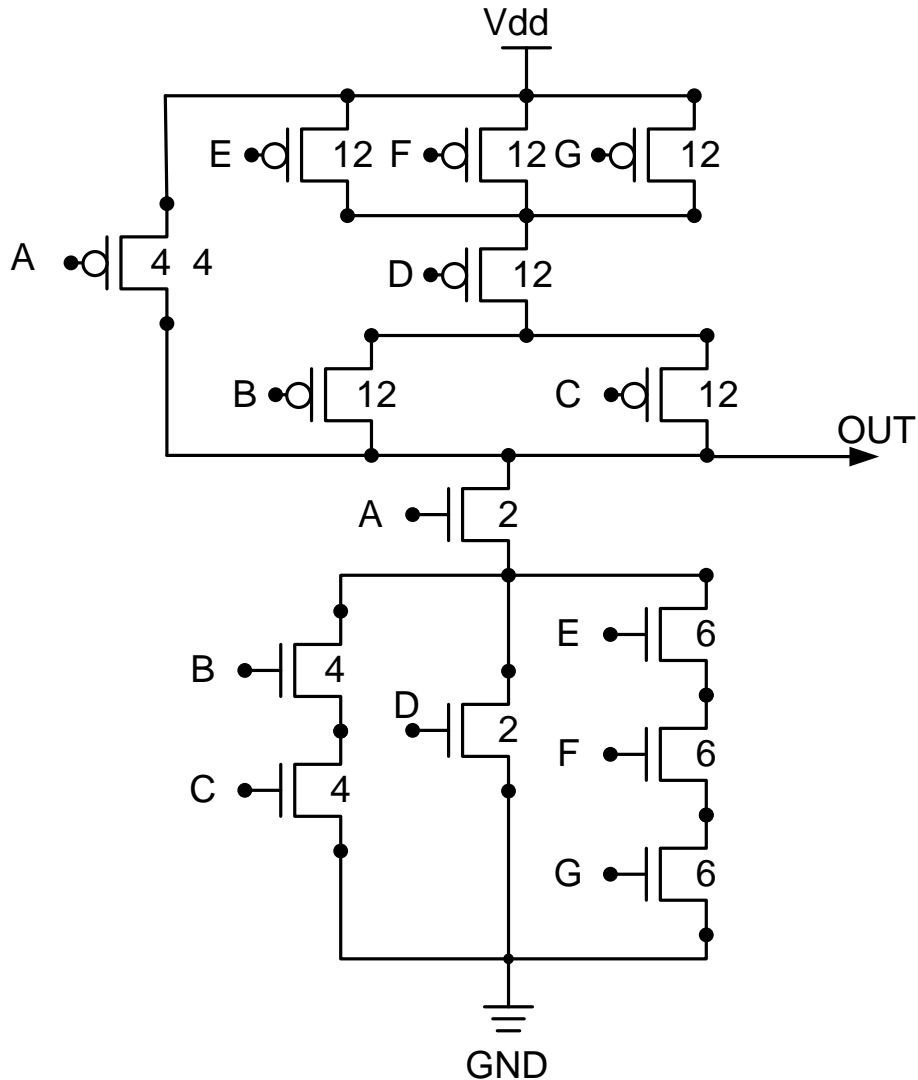
[Helpful Videos to watch for your own knowledge]. Watch the following video clips on CMOS inverter fabrication and also the IC fabrication:

- CMOS Inverter fabrication: [https://www.youtube.com/watch?v=OBiu2agne\\_U](https://www.youtube.com/watch?v=OBiu2agne_U)
- An IC could have billions of transistors inside hundreds of millions of CMOS logic cells. The following video by Intel provides a very high level view of IC fabrication: <https://www.youtube.com/watch?v=aCOyq4YzBtY>
- The following video is based on old technologies, however it still provides some useful information: <https://www.youtube.com/watch?v=gBAKXvsaEiw>
- Interesting video on one application of PN junctions for solar cells. You can watch the video to see how solar energy can create electric power: <https://www.youtube.com/watch?v=ZAxCvHxlLh8>

- Given the following pull-down for a CMOS compound gate.

- Draw the Pull-up network.
- Write the Boolean equation of OUT.
- Go back and show the widths of each NMOS and PMOS transistor such that the current in the worst case for rise time and fall time is equal to that of an inverter with minimum sized NMOS and equal rise and fall time.  $\mu_n/\mu_p = 4$ .

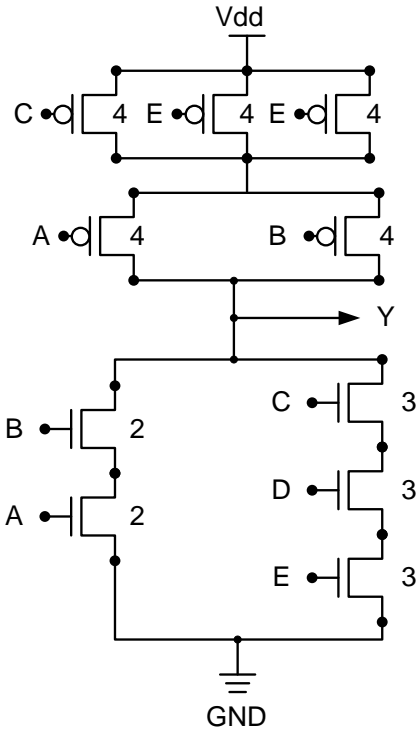




$$\text{OUT} = [A \cdot (BC + D + EFG)]'$$

## 2. CMOS and Sizing

- Draw the transistor level diagram of  $Y = (AB + CDE)'$  using a CMOS approach. Assume the complements (inverse) of each input is NOT available. Try to minimize the number of transistors used.
- Annotate your transistor schematic design from problem 1 to show transistor widths (sizes) so that your implementation of  $Y$  matches the equivalent pull-up and pull-down resistance of a CMOS inverter. Assume the mobility ratio (i.e.  $\mu_n / \mu_p$ ) is 2 and use minimum length transistors ( $L=1$ )

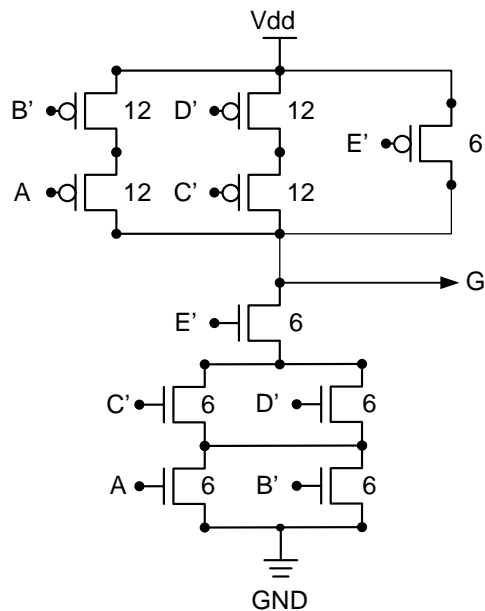


### 3. CMOS and Sizing

- a. Design a CMOS circuit that implements  $G = [A'B + CD + E]$ . Size the transistors such that the CMOS circuit is twice as strong as a reference inverter with NMOS  $W=1$  and PMOS  $W=3$  size. (Assume inputs A, B, C, D and E as well as their complements are available).

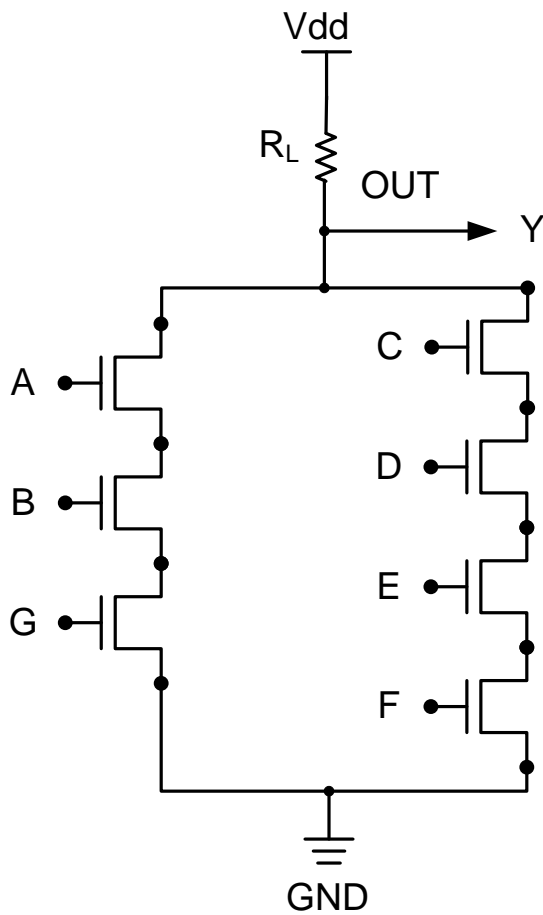
**Note: We want this circuit to be TWICE as strong as an inverter with NMOS  $W=1$  and PMOS  $W=3$ , so we double everything.**

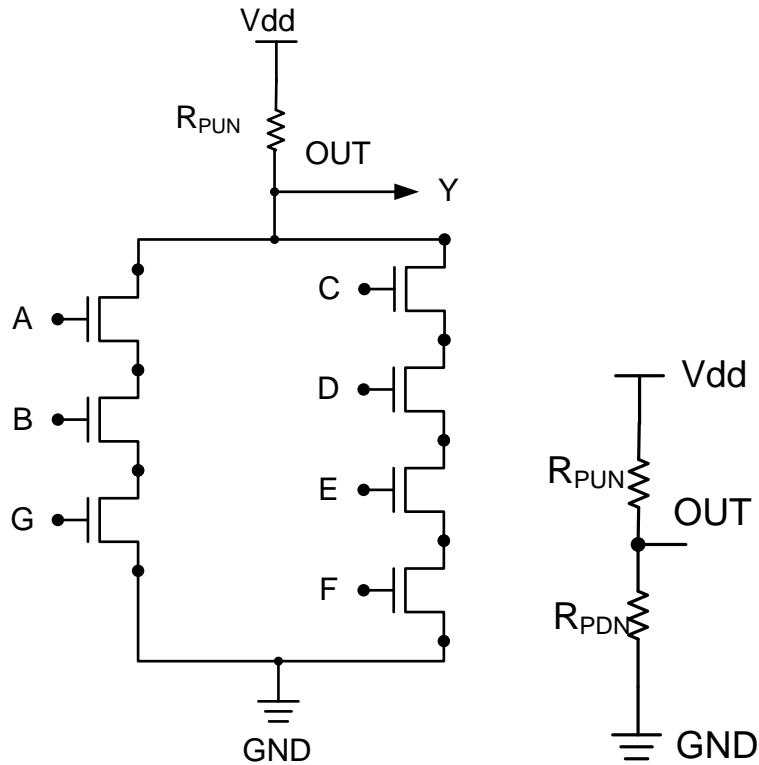
$$G = \overline{[A'B + CD + E]} = \overline{[(A + B')(C' + D')(E')]} = [(A + B')(C' + D')(E')]$$



4. NMOS-only and Resistive Loads

- a. Draw the schematic of  $Y = (ABG + CDEF)'$  using an NMOS-only approach (normal PDN using NMOS transistors but a single resistor in place of the PUN).
- b. Now, assuming that the input combination is such that the output *should* produce a logic '0' and that each NMOS transistor has a resistance of 5Kohm in linear mode, what value resistor would you be needed for the PUN to ensure  $V_{out}$  is at most  $0.25 \cdot V_{dd}$  for the worst case input combination.





If we model the PUN and PDN as single resistors we can use the voltage divider equation:

$$V_{out} = (R_{PDN} / (R_{PDN} + R_{PUN}))V_{dd}$$

The worst case of  $R_{PDN}$  is when it is larger since more voltage will be dropped across it. Thus the worst case is when C,D,E,F = 1,1,1,1 while A,B,G != 1,1,1. If each transistor is modeled as 5Kohms, then we have the equation.

$$0.25V_{dd} = (20K / (20K + R_{PUN}))V_{dd}$$

$$0.25 = (20K / (20K + R_{PUN}))$$

$$5K + 0.25R_{PUN} = 20K$$

$$0.25R_{PUN} = 15K$$

$$R_{PUN} = 60Kohms$$

5. Prove that the unit of RC is seconds.

$$I = Q \text{ (charge)} / T \text{ (seconds)}$$

$$R = V/I$$

$$C = Q / V$$

$$\text{Thus } RC = (V/I) * (Q/V) = Q/I = Q/(Q/T) = T$$

6. RC Transient Analysis: Given the RC circuit on page 9 of unit 18, verify the formulae given on page 11 for the time constant, the propagation delay and the transition time (based on 10%VDD and 90%VDD crossing points) of the output voltage. Show the details of your work.

Use the equation for when a capacitor is discharged and then at time = 0 a voltage source equal to V<sub>dd</sub> is applied:

$$V_c(t) = V_{dd} \left[ 1 - e^{-\frac{t}{RC}} \right]$$

Time constant (i.e. V<sub>c</sub> = 0.63V<sub>dd</sub> when t = RC)

$$0.63 * V_{dd} = V_{dd} \left[ 1 - e^{-\frac{t}{RC}} \right]$$

$$0.37 = e^{-\frac{t}{RC}}$$

$$\ln(0.37) = -\frac{t}{RC}$$

$$RC * -\ln(0.37) = t \approx RC$$

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Propagation delay

$$0.5 * V_{dd} = V_{dd} \left[ 1 - e^{-\frac{t}{RC}} \right]$$

$$\ln(0.5) = -\frac{t}{RC}$$

$$t = 0.69 * RC$$

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Transition time delay (10%-90%) = Time it takes to get to 90% - Time it took to get to 10%

Time it takes to get to 90%

$$0.9 * V_{dd} = V_{dd} \left[ 1 - e^{-\frac{t}{RC}} \right]$$

$$\ln(0.1) = -\frac{t}{RC}$$

$$t = 2.3 * RC$$

Time it takes to get to 10%

$$0.1 * V_{dd} = V_{dd} \left[ 1 - e^{-\frac{t}{RC}} \right]$$

$$\ln(0.9) = -\frac{t}{RC}$$

$$t = 0.1 * RC$$

Transition Time = 2.3RC – 0.1RC = 2.2RC

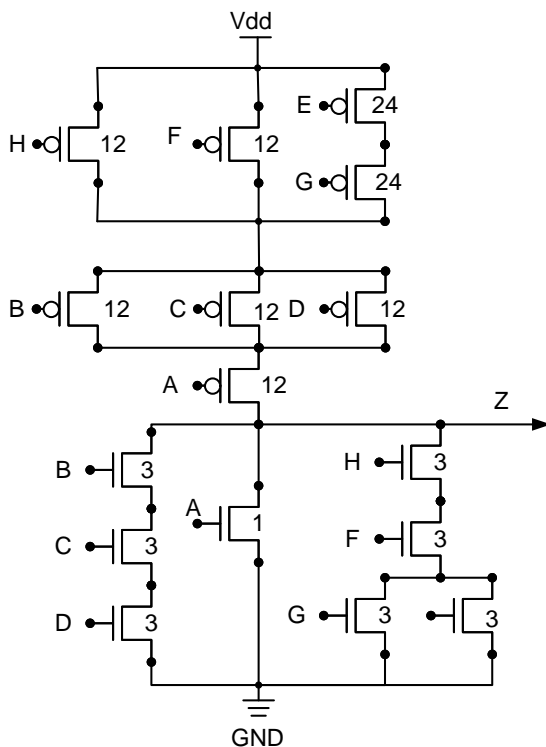
7. Inverter Transient Analysis: The equivalent resistance values of a PMOS and an NMOS in an Inverter are 10KOhms and 7KOhms, respectively. The inverter drives an effective capacitance of 10fF (fF= femtoFarads =  $10^{-15}$  Farads). Ignore other parasitic (internal) capacitances. Using an equivalent RC model to calculate the a.) propagation delay, b.) time constant and c.) transition time (based on 10%VDD and 90%VDD) for BOTH the rising output case and falling output case.

	Rising	Falling
<b>Prop. Delay (0.69RC)</b>	$0.69 * R_p * C = 0.69 * 10K * 10^{-15} = 6.9ps$	$0.69 * R_n * C = 0.69 * 7K * 10^{-15} = 4.83ps$
<b>Time Constant = RC</b>	$R_p * C = 10K * 10^{-15} = 10ps$	$R_n * C = 7K * 10^{-15} = 7ps$
<b>Transition Time (2.2RC)</b>	$2.2 * R_p * C = 2.2 * 10K * 10^{-15} = 22ps$	$2.2 * R_n * C = 2.2 * 7K * 10^{-15} = 15.4ps$

8. CMOS Logic Design and Transient Analysis:

Below is the CMOS logic cell that outputs  $Z = \sim(A+BCD+HF(E+G))$ . The transistors are sized assuming a mobility ratio of 4.

- The equivalent resistance of an NMOS with a size of 1W is measured as 10KOhms and for a PMOS of size 1W is 40KOhms. The effective capacitance of the load is approximated as 100fF. Ignore other parasitic (internal) capacitances. Calculate the propagation delay (time for output to reach 0.5Vdd) of falling output in the WORST CASE scenario. Repeat for the rising output.
- Repeat your calculations to now find the BEST CASE propagation delay of the falling output and rising output



a.)

When the PDN is on we will see a ratio of  $W/L = 1 \Rightarrow 10KOhms$

Prop. Delay (Falling) =  $0.69 * 10Kohms * 100E-15 = 0.69ns$  (690 ps)

When PUN is on, we will see  $W = 12 / L = 3$  for a ratio of 4. Thus, since  $1W/1L = 40Kohms$ , then  $W=12/L=3 \Rightarrow 10KOhms$

Prop. Delay (Rising) =  $0.69 * 10Kohms * 100E-15 = 0.69ns$  (690 ps).

b.) The best case is when all transistors are on. (Recall that resistors in parallel yield  $R_{eq} = (1/R_1 + \dots + 1/R_n)^{-1}$ )

If an NMOS with  $W/L = 1$  yields an equivalent  $10\text{K}\Omega$  resistance, then a  $W/L = 3$  NMOS transistor has resistance of  $10\text{K}/3$ . When B,C,D are on in series we have  $10\text{K}/3 + 10\text{K}/3 + 10\text{K}/3 = 10\text{K}$ . Transistor A is just  $10\text{K}\Omega$  and H,F,E,G yields  $(10\text{K}/3 + 10\text{K}/3 + (10\text{K}/6))$ . So we have a parallel connection of:  $10\text{K} \parallel 10\text{K} \parallel 10\text{K}(5/6)$ .  $= (1/10\text{K} + 1/10\text{K} + 6/50\text{K})^{-1} = (16/50\text{K})^{-1} \Rightarrow$  equivalent  $R_{PDN} = 50\text{K}/16$

Thus fall time is  $3.125\text{K} * 100\text{E}-15 = 312.5 \text{ ps}$

For the PUN the best case is when they are all on. Thus A is  $40\text{K}/12$ . That is in series with  $40\text{K}/(12*3) = 40\text{K}/36$  for B,C,D. The same equivalent resistance ( $40\text{K}/36$ ) is found for H, F, E and G. Putting these in series yield  $R_{PUN} = 40\text{K}/12 + 2*40\text{K}/36 = 50\text{K}/9 = 5.55\text{K}\Omega$

Thus rise time is  $5.55\text{K} * 100\text{E}-15 = 555.5 \text{ ps}$

## 9. CMOS Inverter Static Analysis:

Setup a KCL equation to calculate the output voltage ( $V_{out}$ ) of a CMOS inverter with sizes  $W$  and  $2W$  (for the NMOS and PMOS, respectively) if the input is permanently connected to a high voltage of  $2.4\text{V}$ . Assume  $V_{DD} = 3.0\text{V}$ ,  $V_{tn} = |V_{tp}| = 0.4\text{V}$ ,  $k_n = 2k_p$ . You need only setup the equation and manipulate it to be of the form  $0 = aV_{out}^2 + bV_{out} + c$ . Hint: Use an educated guess for the approximate level of  $V_{out}$  to determine the mode of operation of each transistor.

First, recognize both transistors are on:  $V_{gs,n} > V_{tn}$  ( $2.4 > 0.4$ ) and  $V_{gs,p} < V_{tp}$  ( $-0.6 < -0.4$ ). So to find what mode each transistor is operating we need to "know"  $V_{ds,n}$  and  $V_{ds,p}$ . Since this is an inverter and the input is "high-ish" the output should be "low-ish". If that is the case then  $V_{out}$  is small. Note that  $V_{ds,n} = V_{out}$  and  $V_{ds,p} = V_{DD} - V_{out}$ . So if the output is low-ish we should see that  $V_{gs,n} - V_{tn} > V_{ds,n}$  and thus the NMOS is in LINEAR. For the PMOS,  $|V_{ds,p}| > |V_{gs,p}| - |V_{tp}|$  and thus it is in SATURATION mode. From here we can setup our KCL equation since the current through the PMOS must be the same as through the NMOS.

$$I_{pmos} = I_{nmos}$$

$$0.5 * k_p * 2W * [0.2]^2 = 0.5 * k_n * 1W * [2 * 2.0V * V_{out} - V_{out}^2] =$$

$$0.5 * k_p * 2W * [0.2]^2 = 0.5 * 2 * k_p * 1W * [4.0V * V_{out} - V_{out}^2] =$$

$$0.04 = [4.0V * V_{out} - V_{out}^2]$$

$$1 = [4.0V * V_{out} - V_{out}^2] / 0.04$$

$$0 = -25 * V_{out}^2 + 100 * V_{out} - 1$$